

Multi-D wavelet construction using Quillen-Suslin theorem for Laurent polynomials

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Fourier Talks

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Outline

- 1 Review on Quillen-Suslin theorem and wavelet construction
- 2 Our new approaches for non-redundant wavelet construction

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Quillen-Suslin theorem for Laurent polynomials

A column q -vector $D(z)$ with Laurent polynomial entries is **unimodular** if it has a left inverse, i.e. if there exists a row q -vector $F(z)$ s.t. $F(z)D(z) = 1$.

We assume $z \in \mathbb{C}^n, |z| = 1$.

Example:

$$D(z) = \left[\frac{1}{2}, \frac{1}{4}z_1^{-1} + \frac{1}{4}, \frac{1}{4}z_2^{-1} + \frac{1}{4}, \frac{1}{4}z_1^{-1}z_2^{-1} + \frac{1}{4} \right]^T$$

is unimodular since $[2, 0, 0, 0]$ is a left inverse of $D(z)$.

Another left inverse of $D(z)$ is

$$\left[-\frac{1}{8}z_1^{-1} - \frac{1}{8}z_2^{-1} - \frac{1}{8}z_1^{-1}z_2^{-1} + \frac{5}{4} - \frac{1}{8}z_1 - \frac{1}{8}z_2 - \frac{1}{8}z_1z_2, \frac{1}{4} + \frac{1}{4}z_1, \frac{1}{4} + \frac{1}{4}z_2, \frac{1}{4} + \frac{1}{4}z_1z_2 \right]$$

The first one is simpler but the second one has better accuracy.

Theorem (Quillen-Suslin Thm for Laurent poly by Swan, 1978)

Let $D(z)$ be a unimodular column q -vector. Then there exists an invertible $q \times q$ matrix $T(z)$ s.t. $T(z)D(z) = [1, 0, \dots, 0]^T$.

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Designing filter bank (FB) using Laurent polynomials

Via polyphase representation (Vaidyanathan, 1993)

FB design problem: Find

- $H(z), J_i(z), i = 1, \dots, p - 1$: row q -vectors
- $D(z), K_i(z), i = 1, \dots, p - 1$: column q -vectors s.t.

$$S(z)A(z) := \begin{bmatrix} D(z) & K_1(z) & \cdots & K_{p-1}(z) \end{bmatrix} \begin{bmatrix} H(z) \\ J_1(z) \\ \vdots \\ J_{p-1}(z) \end{bmatrix} = I_q$$

$A(z)$: analysis bank; $S(z)$: synthesis bank.

The above identity is called the perfect reconstruction property.
For the perfect reconstruction property to hold, we need $p \geq q$.

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The FB is called

- a **non-redundant (or biorthogonal) FB** if $p = q$.
In this case, $A(z)$ and $S(z)$ are square matrices.
- a **wavelet FB** if
 - $H(z)$: **lowpass** row q -vector
 - $J_i(z), i = 1, \dots, p - 1$: **highpass** row q -vectors
 - $D(z)$: **lowpass** column q -vector
 - $K_i(z), i = 1, \dots, p - 1$: **highpass** column q -vectors

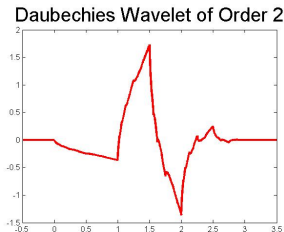
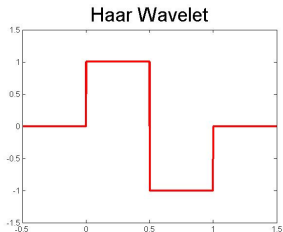
⇒ **wavelet FB design: a key step in wavelet construction**
- a **wavelet FB with m vanishing moments (VM)** (for $m \geq 1$) if
 - $H(z)$: **lowpass** row q -vector
 - $J_i(z), i = 1, \dots, p - 1$: row q -vectors **with m VM**
 - $D(z)$: **lowpass** column q -vector
 - $K_i(z), i = 1, \dots, p - 1$: column q -vectors **with m VM**

⇒ **leads to wavelets with m VM (high performance)**

Very brief introduction to wavelets

- Wavelets are a collection of functions obtained by scaling and translating a fixed set of functions (mother wavelets).
- Wavelet is a subfield of Harmonic analysis and highly interdisciplinary. Wavelets are used in many applications (e.g. image/signal processing, compressive sensing).

Examples (1-D): Haar (1909), VM=1; Daubechies (1987), VM=2

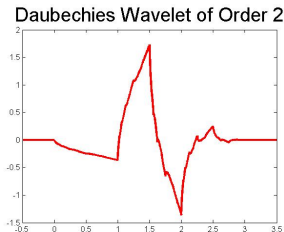
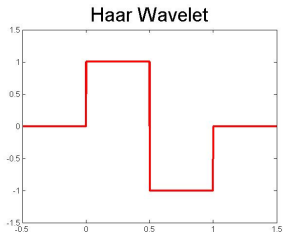


Constructing multi-D wavelets is challenging and important.

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Constructing multi-D wavelets is challenging and important.

Current approach

for designing non-redundant wavelet FBs

Theorem (by Chen-Han-Riemenschneider, 2000)

Suppose $H(z)$, $D(z)$ are lowpass vectors and

$$\begin{bmatrix} D(z) & K_1(z) & \cdots & K_{q-1}(z) \end{bmatrix} \begin{bmatrix} H(z) \\ J_1(z) \\ \vdots \\ J_{q-1}(z) \end{bmatrix} = I_q$$

Then the following are equivalent.

- 1 $H(z)$, $D(z)$ have m accuracy (AC, or approximation order).
- 2 $J_i(z)$, $K_i(z)$, $i = 1, \dots, q - 1$, have m VM.

Corollary (obtained by C-H-R and Q-S for Laurent polynomials)

Let $H(z)$, $D(z)$ be lowpass vectors with m AC and $H(z)D(z) = 1$.
Then there exist $J_i(z)$, $K_i(z)$, $i = 1, \dots, q - 1$, with m VM such that
the perfect reconstruction property holds.

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Vanishing moments (VM) and accuracy (AC)

Assume the dilation is dyadic, and $q = 2^n$.

Then, $\Gamma := \{0, 1\}^n =: \{\nu_0 = 0, \nu_1, \dots, \nu_{q-1}\}$ can be chosen.

Notice that $\mathbb{Z}^n = \cup_{\nu \in \Gamma} (2\mathbb{Z}^n + \nu)$.

Definition (for dyadic dilation case)

For $H(z) = [H_0(z), H_1(z), \dots, H_{q-1}(z)]$, let $H(z) = \sum_{j=0}^{q-1} z^{\nu_j} H_j(z^2)$.
Let m be a nonnegative integer. Then $H(z)$ has

- m VM if $\frac{\partial^k}{\partial \omega^k} H(e^{i\omega})|_{\omega=0} = 0, \forall |k| \leq m - 1$, and
- m AC if $\frac{\partial^k}{\partial \omega^k} H(e^{i\omega})|_{\omega=\gamma} = 0, \forall |k| \leq m - 1, \forall \gamma \in \{0, \pi\}^n \setminus 0$.

VM and AC for column vector $D(z) = [D_0(z), D_1(z), \dots, D_{q-1}(z)]^T$
is defined similarly by forming $D(z) = \sum_{j=0}^{q-1} z^{-\nu_j} D_j(z^2)$.

$H(z)$ or $D(z)$ is the highpass vector iff it has $\text{VM} \geq 1$.

Current approach is not satisfactory

Finding $H(z)$, $D(z)$ satisfying assumptions of Corollary is not easy, especially if the AC or the spatial dimension n is large.

Example: Let $n = 2$, and let

$H(z) = [\frac{1}{2}, \frac{1}{4}z_1^{-1} + \frac{1}{4}, \frac{1}{4}z_2^{-1} + \frac{1}{4}, \frac{1}{4}z_1^{-1}z_2^{-1} + \frac{1}{4}]$: lowpass with 2 AC.
Then $[2, 0, 0, 0]^T$: lowpass, a right inverse of $H(z)$, but with 0 AC.
Using Maple implementation of Algebraic Geometry theory (Cox-Little-O'Shea, 2006), we see any right inverse of $H(z)$ is

$$\begin{bmatrix} 2 \\ 0 \\ 0 \\ 0 \end{bmatrix} - \frac{1}{2}u_1(z) \begin{bmatrix} z_1^{-1} + 1 \\ -2 \\ 0 \\ 0 \end{bmatrix} - \frac{1}{2}u_2(z) \begin{bmatrix} z_2^{-1} + 1 \\ 0 \\ -2 \\ 0 \end{bmatrix} - \frac{1}{2}u_3(z) \begin{bmatrix} z_1^{-1}z_2^{-1} + 1 \\ 0 \\ 0 \\ -2 \end{bmatrix}$$

for some Laurent polynomials $u_1(z), u_2(z), u_3(z)$. To find a right inverse of $H(z)$ with 2 AC, one can use this parameterization. Usually done by fixing the total degree of Laurent poly u_1, u_2, u_3 , and then increasing the total degree if needed (Riemenschneider-Shen, 1997; Han-Jia, 1999; Park, 2002).

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Our approach (theory & algorithm)

for designing non-redundant wavelet FBs

Inputs:

- $H(z)$: row, lowpass, q -vector w/ **unimodularity, positive AC.**
- $G(z)$: column, lowpass, q -vector w/ **positive AC.**
- $F(z)$: column, lowpass, q -vector, **a right inverse of $H(z)$.**

Algorithm: Set

- $D(z) := G(z) + F(z)(1 - H(z)G(z))$: column, q -vector
- $T(z)$: $q \times q$ invertible matrix s.t. $T(z)H(z)^T = [1, 0, \dots, 0]^T$.
- $K_1(z), \dots, K_{q-1}(z)$: 2nd to last columns of $T(z)^T$.
- $J_1(z), \dots, J_{q-1}(z)$: 2nd to last rows of $T(z)^{-T}[I_q - F(z)H(z)][I_q - G(z)H(z)]$.

Output:

Wavelet FB: $(D(z), K_1(z), \dots, K_{q-1}(z)), (H(z), J_1(z), \dots, J_{q-1}(z))$
where $D(z)$ is a right inverse of $H(z)$ with positive AC.

Example: Designing a 2-D wavelet FB

Let $n = 2$.

Let $H(z) = G(z)^* = [\frac{1}{2}, \frac{1}{4}z_1^{-1} + \frac{1}{4}, \frac{1}{4}z_2^{-1} + \frac{1}{4}, \frac{1}{4}z_1^{-1}z_2^{-1} + \frac{1}{4}]$: lowpass with 2 AC.

Let $F(z) = [2, 0, 0, 0]^T$: lowpass with $H(z)F(z) = 1$, but with 0 AC.

Set $D(z) = G(z) + F(z)(1 - H(z)G(z)) = H(z)^* + (1 - H(z)H(z)^*)F(z)$

$$= \begin{bmatrix} \frac{1}{2} \\ \frac{1}{4} + \frac{1}{4}z_1 \\ \frac{1}{4} + \frac{1}{4}z_2 \\ \frac{1}{4} + \frac{1}{4}z_1z_2 \end{bmatrix} + \left(-\frac{1}{16}z_1^{-1} - \frac{1}{16}z_2^{-1} - \frac{1}{16}z_1^{-1}z_2^{-1} + \frac{3}{8} - \frac{1}{16}z_1 - \frac{1}{16}z_2 - \frac{1}{16}z_1z_2 \right) \begin{bmatrix} 2 \\ 0 \\ 0 \\ 0 \end{bmatrix}$$

$$= \left[-\frac{1}{8}z_1^{-1} - \frac{1}{8}z_2^{-1} - \frac{1}{8}z_1^{-1}z_2^{-1} + \frac{5}{4} - \frac{1}{8}z_1 - \frac{1}{8}z_2 - \frac{1}{8}z_1z_2, \frac{1}{4} + \frac{1}{4}z_1, \frac{1}{4} + \frac{1}{4}z_2, \frac{1}{4} + \frac{1}{4}z_1z_2 \right]^T$$

We see that $D(z)$ has 2 AC.

From the implementation of Quillen-Suslin Theorem by Maple, we see that

$$T(z) := \begin{bmatrix} 2 & 0 & 0 & 0 \\ -\frac{1}{2}z_1^{-1} - \frac{1}{2} & 1 & 0 & 0 \\ -\frac{1}{2}z_2^{-1} - \frac{1}{2} & 0 & 1 & 0 \\ -\frac{1}{2}z_1^{-1}z_2^{-1} - \frac{1}{2} & 0 & 0 & 1 \end{bmatrix} \text{ satisfies } T(z)H(z)^T = \begin{bmatrix} 1 \\ 0 \\ 0 \\ 0 \end{bmatrix}.$$

Hence, we can get $K_1(z), K_2(z), K_3(z)$ from 2nd to 4th columns of $T(z)^T$ and

$J_1(z), J_2(z), J_3(z)$ from 2nd to 4th rows of $T(z)^{-T}[I_4 - F(z)H(z)][I_4 - G(z)H(z)]$.

Example: Designing a 2-D wavelet FB

More precisely, the analysis bank $A(z)$ and the synthesis bank $S(z)$ are given as

$$A(z) = \begin{bmatrix} H(z) \\ J_1(z) \\ J_2(z) \\ J_3(z) \end{bmatrix}, \quad S(z) = [D(z) \quad K_1(z) \quad K_2(z) \quad K_3(z)]$$

where

$$\begin{aligned} J_1(z) &= \begin{bmatrix} -\frac{1}{8} - \frac{1}{8}z_1 & -\frac{1}{16}z_1^{-1} + \frac{7}{8} - \frac{1}{16}z_1 & -\frac{1}{16}z_2^{-1} - \frac{1}{16}z_2^{-1}z_1 - \frac{1}{16} - \frac{1}{16}z_1 & -\frac{1}{16}z_1^{-1}z_2^{-1} - \frac{1}{16}z_2^{-1} - \frac{1}{16} - \frac{1}{16}z_1 \end{bmatrix} \\ J_2(z) &= \begin{bmatrix} -\frac{1}{8} - \frac{1}{8}z_2 & -\frac{1}{16}z_1^{-1} - \frac{1}{16}z_1^{-1}z_2 - \frac{1}{16} - \frac{1}{16}z_2 & -\frac{1}{16}z_2^{-1} + \frac{7}{8} - \frac{1}{16}z_2 & -\frac{1}{16}z_1^{-1}z_2^{-1} - \frac{1}{16}z_1^{-1} - \frac{1}{16} - \frac{1}{16}z_2 \end{bmatrix} \\ J_3(z) &= \begin{bmatrix} -\frac{1}{8} - \frac{1}{8}z_1z_2 & -\frac{1}{16}z_1^{-1} - \frac{1}{16} - \frac{1}{16}z_2 - \frac{1}{16}z_1z_2 & -\frac{1}{16}z_2^{-1} - \frac{1}{16} - \frac{1}{16}z_1 - \frac{1}{16}z_1z_2 & -\frac{1}{16}z_1^{-1}z_2^{-1} + \frac{7}{8} - \frac{1}{16}z_1z_2 \end{bmatrix} \end{aligned}$$

and

$$K_1(z) = \begin{bmatrix} -\frac{1}{2}z_1^{-1} - \frac{1}{2} \\ 1 \\ 0 \\ 0 \end{bmatrix} \quad K_2(z) = \begin{bmatrix} -\frac{1}{2}z_2^{-1} - \frac{1}{2} \\ 0 \\ 1 \\ 0 \end{bmatrix} \quad K_3(z) = \begin{bmatrix} -\frac{1}{2}z_1^{-1}z_2^{-1} - \frac{1}{2} \\ 0 \\ 0 \\ 1 \end{bmatrix}$$

Our approach for high-performance wavelet FB design

Given a positive integer m , let

- $H(z)$: row, lowpass, q -vector w/ **unimodularity**, m AC.
- $G(z)$: column, lowpass, q -vector w/ m AC.
- $F(z)$: column, lowpass, q -vector, **a right inverse of $H(z)$** .

Then **one has an algorithm to construct a non-redundant wavelet FB with at least m VM** so that

$$\begin{bmatrix} D(z) & * & \cdots & * \end{bmatrix} \begin{bmatrix} H(z) \\ * \\ \vdots \\ * \end{bmatrix} = \mathbb{I}_q,$$

where $D(z)$, still determined by $H(z)$, $G(z)$ and $F(z)$, is a right inverse of $H(z)$ with at least m AC.

Summary

Summary of the talk

- Our method can be used to design non-redundant wavelet FBs for any dimension (it is especially useful for multi-D).
- It provides an algorithm for constructing a wavelet FB w/ m VM, starting from a unimodular lowpass vector w/ m AC.
- It does not require the initial unimodular lowpass vector to satisfy any additional assumption other than AC condition.

Things that I did not cover in the talk

- Our approaches work for any dilation.
- Connection of our method to Laplacian pyramid algorithms.

Remaining challenges

- Our approaches so far have been mostly algebraic, hence questions that are analytic in nature need to be answered separately.
- Currently we are using only the implementation of the big theorem (Quillen-Suslin Theorem for Laurent polynomials). We'll try to fully exploit the powerfulness of the big theorem.
- Currently we are concerned with only the VM of wavelets. We'll try to incorporate other properties such as symmetry, interpolatory property, and fast algorithms.

References

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Thank you for your attention

Outline

3 Appendix

Unimodularity is ring-dependent

$[z, z^2]$ is unimodular in Laurent polynomial ring since $[\frac{1}{2}z^{-1}, \frac{1}{2}z^{-2}]^T$ is a right inverse, but not in polynomial ring since there are no polynomials $f(z), g(z)$ s.t. $f(z)z + g(z)z^2 = 1$.

Filter bank (FB)

$h, d, j_i, k_i : \mathbb{Z}^n \rightarrow \mathbb{R}$, $i = 1, \dots, p-1$, are filters (w/ finite supports)
 Downsampling & upsampling, with $n \times n$ sampling matrix Λ
 (w/ integers and all eigenvalues have magnitude larger than 1):

$$y_{\downarrow}(k) = y(\Lambda k), \quad k \in \mathbb{Z}^n.$$

$$y_{\uparrow}(k) = \begin{cases} y(\Lambda^{-1}k), & k \in \Lambda\mathbb{Z}^n, \\ 0, & \text{otherwise.} \end{cases}$$

i.e. for $n = 1$, $\Lambda = 2$, $y = (\dots, y(-1), y(0), y(1), \dots)$, we have

$$y_{\downarrow} = (\dots, y(-2), y(0), y(2), \dots)$$

$$y_{\uparrow} = (\dots, y(-1), 0, y(0), 0, y(1), \dots).$$

Filter bank (FB) problem is to find $\{h, j_1, \dots, j_{p-1}\}$, $\{d, k_1, \dots, k_{p-1}\}$
 s.t. $d * ((h * x)_{\downarrow})_{\uparrow} + \sum_{i=1}^{p-1} k_i * ((j_i * x)_{\downarrow})_{\uparrow} = x$, for any finitely
 supported signal x .

Laurent polynomial lowpass/highpass vectors

Λ : $n \times n$ sampling matrix

$$q = |\det \Lambda|$$

Γ : a set of representatives of distinct cosets of $\mathbb{Z}^n / \Lambda\mathbb{Z}^n$ with 0.

Then $\Gamma =: \{\nu_0 = 0, \nu_1, \dots, \nu_{q-1}\}$, $\mathbb{Z}^n = \cup_{\nu \in \Gamma} (\Lambda\mathbb{Z}^n + \nu)$.

Definition

Let $H(z) = [H_0(z), H_1(z), \dots, H_{q-1}(z)]$ be a row q -vector.

- $H(z)$ is the (polyphase) **lowpass** vector if $H(e^{i\omega})|_{\omega=0} = \sqrt{q}$
- $H(z)$ is the (polyphase) **highpass** vector if $H(e^{i\omega})|_{\omega=0} = 0$

where

$$H(z) = \sum_{j=0}^{q-1} z^{\nu_j} H_j(z^\Lambda)$$

The type of column q -vector $D(z) = [D_0(z), D_1(z), \dots, D_{q-1}(z)]^T$ is defined similarly by forming $D(z) = \sum_{j=0}^{q-1} z^{-\nu_j} D_j(z^\Lambda)$.

Our theory: full version

Theorem (by Hur-Park-Zheng)

Let

- $\alpha_H, \beta_H, \alpha_G, \beta_G > 0$ and $\alpha_F \geq 0$: integers.
- $H(z)$: unimodular lowpass row q -vector w/ α_H AC, β_H FL.
- $G(z)$: lowpass column q -vector w/ α_G AC and β_G FL.
- $F(z)$: lowpass column q -vector w/ α_F AC, $H(z)F(z) = 1$.

Then one can construct a non-redundant wavelet FB so that

$$\begin{bmatrix} D(z) & * & \cdots & * \end{bmatrix} \begin{bmatrix} H(z) \\ * \\ \vdots \\ * \end{bmatrix} = I_q, D(z) := G(z) + F(z)(1 - H(z)G(z))$$

with lowpass $D(z)$ w/ at least $\min\{\alpha_G, \alpha_F + \beta_G, \alpha_F + \beta_H\} > 0$ AC.

FL: flatness

Our algorithm to get a wavelet FB: full version

Inputs:

- $\alpha_H, \beta_H, \alpha_G, \beta_G > 0$ and $\alpha_F \geq 0$: integers.
- $H(z)$: unimodular lowpass row q -vector w/ α_H AC, β_H FL.
- $G(z)$: lowpass column q -vector w/ α_G AC and β_G FL.
- $F(z)$: lowpass column q -vector w/ α_F AC, $H(z)F(z) = 1$.

Output:

- wavelet FB whose lowpass row vector is $H(z)$.

Procedure:

Step 1 Set $D(z) := G(z) + F(z)(1 - H(z)G(z))$.

Step 2 Find invertible $K(z)$ s.t. $K(z)H(z)^T = [1, 0, \dots, 0]^T$.

Step 3 Let $K_1(z), \dots, K_{q-1}(z)$ be the 2nd to last columns of $K(z)^T$.

Step 4 Let $J_1(z), \dots, J_{q-1}(z)$ be the 2nd to last rows of $K(z)^{-T} [I_q - F(z)H(z)] [I_q - G(z)H(z)]$

Our algorithm to get a wavelet FB w/ *at least* α_H VM: full version

Inputs:

- $\alpha_H, \beta_H > 0$ and $\alpha_F \geq 0$: integers.
- $H(z)$: unimodular lowpass row q -vector w/ α_H AC, β_H FL.
- $F(z)$: lowpass column q -vector w/ α_F AC, $H(z)F(z) = 1$.

Output:

- wavelet FB w/ lowpass row vector $H(z)$ and at least α_H VM.

Procedure:

Step 1 Initialize $Iter := 1$ and $D(z) := H(z)^* + F(z)(1 - H(z)H(z)^*)$

Step 2 While $(\alpha_F + (Iter)\beta_H) < \alpha_H$

$Iter := Iter + 1$; $D(z) := H(z)^* + D(z)(1 - H(z)H(z)^*)$

end

Step 3 Find invertible $K(z)$ s.t. $K(z)H(z)^T = [1, 0, \dots, 0]^T$.

Step 4 Define $K_1(z), \dots, K_{q-1}(z)$ and $J_1(z), \dots, J_{q-1}(z)$ as previous.

Vanishing moments (VM), flatness (FL), accuracy (AC)

Definition

Definition

For $H(z) = [H_0(z), H_1(z), \dots, H_{q-1}(z)]$, let $H(z) = \sum_{j=0}^{q-1} z^{\nu_j} H_j(z^\Lambda)$. Let m be a nonnegative integer.

- $H(z)$ has m **VM** if $\frac{d^k}{d\omega^k} H(e^{i\omega})|_{\omega=0} = 0, \forall |k| \leq m - 1$
- $H(z)$ has m **FL** if $\frac{d^k}{d\omega^k} (\sqrt{q} - H(e^{i\omega}))|_{\omega=0} = 0, \forall |k| \leq m - 1$
- $H(z)$ has m **AC** if $\frac{d^k}{d\omega^k} H(e^{i\omega})|_{\omega=\gamma} = 0, \forall |k| \leq m - 1, \forall \gamma \in \Gamma^* \setminus 0$
 Γ^* : set of rep. of distinct cosets of $2\pi((\Lambda^T)^{-1}\mathbb{Z}^n)/\mathbb{Z}^n$ w/ 0.

VM, FL, AC for column vector $D(z) = [D_0(z), D_1(z), \dots, D_{q-1}(z)]^T$ is defined similarly by forming $D(z) = \sum_{j=0}^{q-1} z^{-\nu_j} D_j(z^\Lambda)$.

$H(z)$ or $D(z)$ is the highpass vector iff it has $\text{VM} \geq 1$

$H(z)$ or $D(z)$ is the lowpass vector iff it has $\text{FL} \geq 1$

Equivalent conditions for VM, FL, and AC

Theorem

For $H(z) = [H_0(z), H_1(z), \dots, H_{q-1}(z)]$, let $H(z) = \sum_{j=0}^{q-1} z^{\nu_j} H_j(z^\Lambda)$.
Let m be a nonnegative integer.

- $H(z)$ has m **VM** iff $H(e^{i\omega}) \approx O(|\omega|^m)$ (at $\omega = 0$)
- $H(z)$ has m **FL** iff $\sqrt{q} - H(e^{i\omega}) \approx O(|\omega|^m)$ (at $\omega = 0$)
- $H(z)$ has m **AC** iff $H(e^{i(\omega+\gamma)}) \approx O(|\omega|^m)$ (at $\omega = 0$), $\forall \gamma \in \Gamma^* \setminus 0$
 Γ^* : set of rep. of distinct cosets of $2\pi((\Lambda^T)^{-1}\mathbb{Z}^n)/\mathbb{Z}^n$ w/ 0.